Optimal modal Fourier transform wave-front Control

Lisa A. Poyneer and Jean-Pierre Véran Lawrence Livermore National Lab, Herzberg Institute of Astrophysics



UCRL-PRES-207798



National Research Council Canada

Conseil national de recherches Canada

This work was performed under the auspices of the U.S. Department of Energy by the University of California, Lawrence Livermore National Laboratory under contract No. W-7405-Eng-48.

High-precision wave-front control

Adapt to changing conditions

High spatial resolution

High frame-rates for control system

1tmosoherickurbulence



Lisa A. Poyneer's presentation on Optimal Fourier Control





FTR modes are sines and cosines

§ FTR uses the DFT in the filtering process

$$X[k,l] = \frac{1}{N} \sum_{m=0}^{N-1} \sum_{n=0}^{N-1} x[m,n] \exp\left(\frac{-j2\pi(km+ln)}{N}\right)$$

§ Modal coefficients are obtainable directly from the DFT values

$$\langle x[m,n], \mathcal{C}_{k,l}[m,n] \rangle = \frac{1}{D_{k,l}} \operatorname{Re} \left\{ X[k,l] \right\}$$
$$\langle x[m,n], \mathcal{S}_{k,l}[m,n] \rangle = \frac{-1}{D_{k,l}} \operatorname{Im} \left\{ X[k,l] \right\}$$

Lisa A. Poyneer's presentation on **Optimal Fourier Control**

Modes are eigenfunctions

- § Fourier modes are eigenfunctions of linear, shift-invariant (LSI) systems
 - **§** The modes for the slopes (on a square aperture) are the same as the modes for the phase
 - § A cosine of phase at frequency [k,l] produces x- and y-slopes only at the cosine and sine of that frequency [k,l]
 - § Where Mx[k,I] describes the filter which measures the x-slopes from actuator commands

 $\begin{array}{ll} \begin{array}{l} \textbf{phase} & \textbf{x-slope} \\ \mathcal{C}_{k,l}[m,n] & \cdot & A\mathcal{C}_{k,l}[m,n] + B\mathcal{S}_{k,l}[m,n] \\ S_{k,l}[m,n] & \cdot & -B\mathcal{C}_{k,l}[m,n] + A\mathcal{S}_{k,l}[m,n] \\ A = \operatorname{Re}\{M_x[k,l]\}, \ B = -\operatorname{Im}\{M_x[k,l]\} \end{array}$

Optimal modal control scheme

- § We follow Altair's implementation and assume an approximate model of control system (exact in simulation case) for each of the independent modes.
- § We control a mode with feedback in the presence of noise.



Block diagram of control loop for a modal coefficient

Optimize the squared-residual error

- § Since the noise at any step is independent of past errors, if we minimize on the measurement s, we minimize on the residual error.
- § If we had perfect knowledge we would minimize

$$\mathcal{J} = \int \left| \frac{1}{1 + \exp(-j\omega)H(\omega)} \right|^2 \left[M(\omega) + N(\omega) \right] \, d\omega$$

§ But we don't... so we have to estimate the open-loop PSD from the closed-loop measurements using

$$\hat{M}(\omega) + \hat{N}(\omega) = \left|1 + \exp(-j\omega)H_0(\omega)\right|^2 \hat{S}(\omega)$$



Modes correspond to PSF locations

- § Each Fourier mode lives at a specific spatial frequency pair [k,l]
- Because the PSF is approximately the PSD of the residual phase (to second order), each Fourier mode appears at a specific location in the PSF



Black = 0, White = 1

Significant reduction in residual error



N=48, NGS Mag 8 example for 8 iterations of gain optimization

- § Use of optimal gains improves performance
 - § significant reduction in residual MSE at each timestep
 - § less variation in MSE at each timestep

Contrast improved in PSF

- § N=48 case with WFS SNR of 2.16
- § Strehl increased from 0.75 to 0.87 (+12%)
- § MSE in band reduced from 0.224 to 0.074 (3 times less)







Lisa A. Poyneer's presentation on **Optimal Fourier Control**

Optimal filter reflects conditions

- § Input phase aberration is a frozen sheet of phase moving across the aperture
- § Deformable Mirror (DM) has unknown low-pass response which attenuates high spatial frequencies
- § Optimal gains compensate for both



Example filter, N=64

Trade bandwidth and sensor errors



Data for N=48, median over a set of 25 random phase screens

- § At high SNRs, optimal gains produce equivalent or more measurement error but less temporal error than before
- § At low SNRs, optimal gains produce less measurement error but more temporal error than before

Computational load is satisfiable today

- § FTR each timestep: $15N^2 \lg N + 20N^2$
- § Estimating periodograms for t steps of telemetry:

 $N^2(5+2.5\lg t)$

§ Averaging the periodograms and finding the optimal gain (k is for evaluations in root-finding):

 $N^2(1+k) + 4k$

§ Assuming k = 10 (using fast method), a 64x64 system at 2.5k kHz has a maximum load of 1.43 GFLOPs/sec.

OFC is both fast and smart

- § Optimal Fourier Control combines the best of state-of-the art approaches:
 - § computationally efficient
 - *§* adaptive control optimizes performance given current observing conditions
- § Further research areas
 - **§** sensitivity analysis to determine performance across a wide range of conditions
 - § exploration of more complex control laws or predictive control